# **RC1587**

# 3A Adjustable/Fixed Low Dropout Linear Regulator

#### **Features**

- · Fast transient response
- Low dropout voltage at up to 3A
- Load regulation: 0.05% typical
- Total output accuracy ±0.5% typical
- Trimmed current limit
- · On-chip thermal limiting
- Standard TO-220 and TO-263 packages

### **Applications**

- Pentium<sup>®</sup> Pro and Pentium II GTL+ bus supply
- · Low voltage logic supply
- · Battery-powered circuitry
- · Post regulator for switching supply

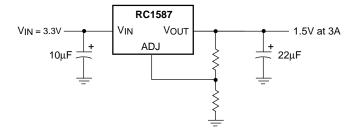
### **Description**

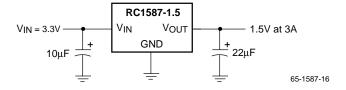
The RC1587 and RC1587-1.5 are low dropout three-terminal regulators with 3A output current capability. These devices have been optimized for low voltage applications including VTT bus termination, where transient response and minimum input voltage are critical. The RC1587 is ideal for low voltage microprocessor applications requiring a regulated output from 1.5V to 3.6V with an input supply of 5V or less. The RC1587-1.5 offers fixed 1.5V with 3A current capability for GTL+ bus VTT termination.

Current limit is trimmed to ensure specified output current and controlled short-circuit current. On-chip thermal limiting provides protection against any combination of overload that would create excessive junction temperatures.

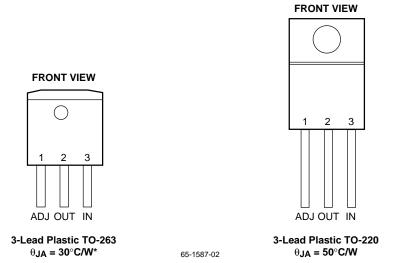
The RC1587 and RC1587-1.5 are available in the industry-standard TO-220 and TO-263 power packages.

### **Block Diagram**





### **Pin Assignments**



\*With package soldered to 0.5 square inch copper area over backside ground plane or internal power plane.  $\theta_{JA}$  can vary from 20°C/W to >40°C/W with other mounting techniques.

## **Absolute Maximum Ratings**

Parameter	Min.	Max.	Unit
VIN		7	V
Operating Junction Temperature Range			
Control Section	0	125	°C
Power Transistor	0	150	°C
Storage Temperature Range	-65	150	°C
Lead Temperature (Soldering, 10 sec.)		300	°C

### **Electrical Characteristics**

Preconditioning: 100% Thermal Limit Functional Test. Tj = 25°C unless otherwise specified.

The • denotes specifications which apply over the specified operating temperature range.

Parameter	Conditions		Min.	Тур.	Max	Units
Reference Voltage	$1.5V \le (V_{IN} - V_{OUT}) \le 5.75V,$ $10mA \le I_{OUT} \le 3A$	•	1.225 (-2%)	1.250	1.275 (+2%)	V
Output Voltage RC1587-1.5	$\begin{aligned} \text{VIN} &= 3.3 \text{V} \pm 0.1 \\ \text{0mA} &\leq \text{IOUT} \leq 3 \text{A} \end{aligned}$	•	1.47	1.5	1.53	V
Line Regulation <sup>1, 2</sup>	$2.75V \le V_{IN} \le 5.75V$ , $I_{OUT} = 10mA$			0.005	0.2	%
Load Regulation <sup>1, 2, 3</sup>	(VIN - VOUT) = 3V, TJ = 25°C, $10mA \le IOUT \le 3A$			0.05	0.3	%
Dropout Voltage	ΔVREF = 1%, I <sub>OUT</sub> = 3A			1.150	1.300	V
Current Limit <sup>3</sup>	(VIN – VOUT) = 5.5V		3.1	3.75		Α
Adjust Pin Current		•		35	120	μΑ
Adjust Pin Current Range	$1.5V \le (V_{IN} - V_{OUT}) \le 5.75V$ , $10mA \le I_{OUT} \le 3A$	•		0.2	5	μΑ
Minimum Load Current	1.5V ≤ (VIN - VOUT) ≤ 5.75V	•		2	10	mA
Quiescent Current	VIN = 5V	•		8	13	mA
Ripple Rejection	$f = 120$ Hz, $C_{OUT} = 25\mu F$ Tantalum, $(V_{IN} - V_{OUT}) = 3V$ , $I_{OUT} = 3A$		60	72		dB
Thermal Regulation	T <sub>A</sub> = 25°C, 30ms pulse			0.004	0.02	%/W
Temperature Stability		•		0.5		%
Long-Term Stability	T <sub>A</sub> = 125°C, 1000 hrs.			0.03	1.0	%
RMS Output Noise (% of VOUT)	TA = 25°C, 10Hz ≤ f ≤ 10kHz			0.003		%
Thermal Resistance,	Control Section				0.7	°C/W
Junction to Case	Power Transistor				3.0	°C/W

- 1. See thermal regulation specifications for changes in output voltage due to heating effects. Load and line regulation are measured at a constant junction temperature by low duty cycle pulse testing.
- 2. Line and load regulation are guaranteed up to the maximum power dissipation (18W for the RC1587). Power dissipation is determined by input/output differential and the output currrent. Guaranteed maximum output power will not be available over the full input/output voltage range.
- 3. The RC1587 has a constant current limit with changes in input-to-output voltage.

## **Typical Performance Characteristics**

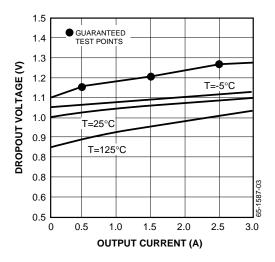


Figure 1. Dropout Voltage vs. Output Current

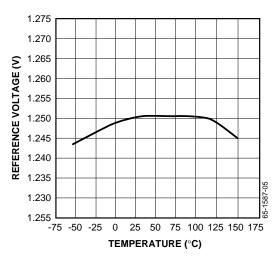


Figure 3. Reference Voltage vs. Temperature

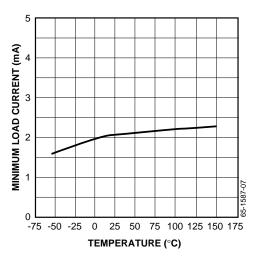


Figure 5. Minimum Load Current vs. Temperature

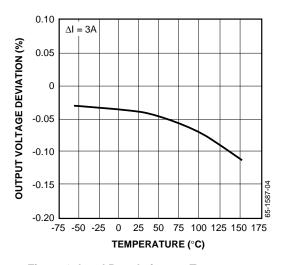


Figure 2. Load Regulation vs. Temperature

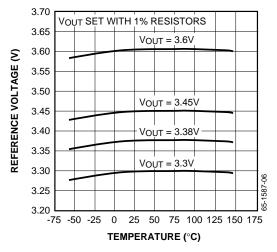


Figure 4. Output Voltage vs. Temperature

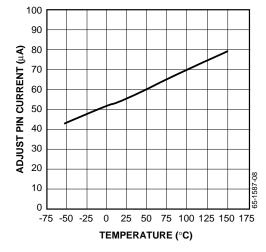
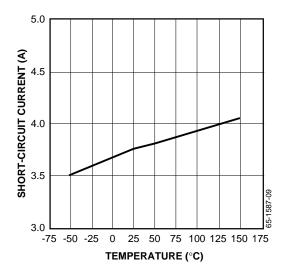


Figure 6. Adjust Pin Current vs. Temperature

### Typical Performance Characteristics (continued)



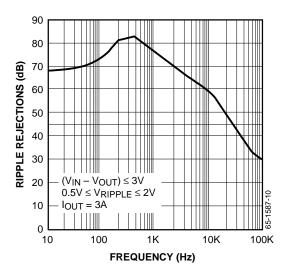


Figure 7. Short-Circuit Current vs. Temperature

Figure 8. Ripple Rejection vs. Frequency

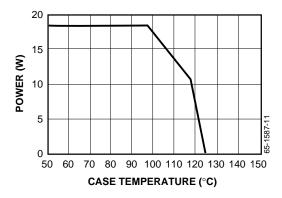


Figure 9. Maximum Power Dissipation

### **Applications Information**

#### General

The RC1587 and RC1587-1.5 are three-terminal regulators optimized for GTL+ VTT termination applications. These devices are short-circuit protected, safe area protected, and offer thermal shutdown to turn off the regulator when the junction temperature exceeds about 150°C. The RC1587 and RC1587-1.5 provide low dropout voltage and fast transient response. Frequency compensation uses capacitors with low ESR while still maintaining stability. This is critical in addressing the needs of low voltage high speed microprocessor buses like GTL+.

#### **Stability**

The RC1587 and RC1587-1.5 require an output capacitor as a part of the frequency compensation. It is recommended to use a  $22\mu F$  solid tantalum or a  $100\mu F$  aluminum electrolytic

on the output to ensure stability. The frequency compensation of these devices optimize frequency response with low ESR capacitors. In general, it is suggested to use capacitors with an ESR of  $<1\Omega$ . It is also recommended to use bypass capacitors such as a 22 $\mu$ F tantalum or a 100 $\mu$ F aluminum on the adjust pin of the devices for low ripple and fast transient response. When these bypassing capacitors are not used at adjust pin, smaller values of output capacitors provide equally good results.

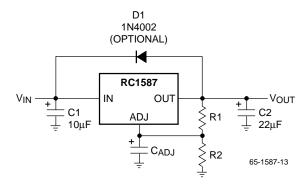
#### **Protection Diodes**

In normal operation, the RC1587 does not require any protection diodes. Internal resistors limit internal current paths on the adjust pin. Therefore, even with bypass capacitors on the adjust pin, no protection diode is needed to ensure device safety under shortcircuit conditions.

A protection diode between the input and output pins is usually not needed. An internal diode between the input and out-

put pins on the RC1587 can handle microsecond surge currents of 50A to 100A. Even with large value output capacitors it is difficult to obtain those values of surge currents in normal operation. Only with large values of output capacitance, such as  $1000\mu F$  to  $5000\mu F$ , and with the input pin instantaneously shorted to ground can damage occur. A crowbar circuit at the input of the RC1587 can generate those levels of current; a diode from output to input is then recommended, as shown in Figure 11. Usually, normal power supply cycling or system "hot plugging and unplugging" will not generate current large enough to do any damage.

The adjust pin can be driven on a transient basis  $\pm 7V$  with respect to the output, without any device degradation. As with any IC regulator, exceeding the maximum input-to-output voltage differential causes the internal transistors to break down and none of the protection circuitry is then functional.



**Figure 11. Optional Protection Diode** 

#### **Ripple Rejection**

In applications that require improved ripple rejection, a bypass capacitor from the adjust pin to ground reduces the output ripple by the ratio of VOUT/1.25V. The impedance of the adjust pin capacitor at the ripple frequency should be less than the value of R1 (typically in the range of  $100\Omega$  to  $120\Omega$ ) in the feedback divider network in Figure 11. Therefore, the value of the required adjust pin capacitor is a function of the input ripple frequency. For example, if R1 equals  $100\Omega$  and the ripple frequency equals 120Hz, the adjust pin capacitor should be 22pF. At 10kHz, only 0.22µF is needed.

#### **Output Voltage**

The RC1587 regulator develops a 1.25V reference voltage between the output pin and the adjust pin (see Figure 12). Placing a resistor R1 between these two terminals causes a constant current to flow through R1 and down through R2 to set the overall output voltage. Normally, this current is the specified minimum load current of 10mA. The current out of the adjust pin adds to the current from R1 and is typically  $55\mu A$ . Its output voltage contribution is small and only needs consideration when very precise output voltage setting is required.

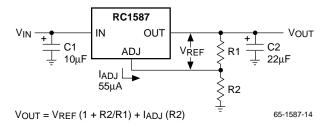


Figure 12. Basic Regulator Circuit

#### **Load Regulation**

It is not possible to provide true remote load sensing because the RC1587 is a three-terminal device. Load regulation is limited by the resistance of the wire connecting the regulators to the load. Load regulation per the data sheet specification is measured at the bottom of the package.

For fixed voltage devices, negative side sensing is a true Kelvin connection with the ground pin of the device returned to the negative side of the load. This is illustrated in Figure 13.

Negative side sensing is a true Kelvin connection with the bottom of the output divider returned to the negative side of the load. The best load regulation is obtained when the top of resistor divider R1 connects directly to the regulator output and not to the load. Figure 13 illustrates this point. If R1 connects to the load, the effective resistance between the regulator and the load is:

$$RP \times (1 + R2/R1)$$
,  $RP = Parasitic Line Resistance$ 

The connection shown in Figure 13 does not multiply Rp by the divider ratio. As an example, Rp is about four milliohms per foot with 16-gauge wire. This translates to 4mV per foot at 1A load current. At higher load currents, this drop represents a significant percentage of the overall regulation. It is important to keep the positive lead between the regulator and the load as short as possible and to use large wire or PC board traces.

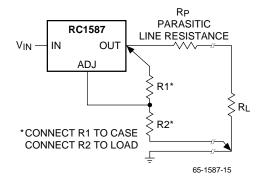


Figure 13. Connection for Best Load Regulation

#### **Thermal Considerations**

The RC1587 protects the device under overload conditions with internal power and thermal limiting circuitry. However, for normal continuous load conditions, do not exceed maximum junction temperature ratings. It is important to consider all sources of thermal resistance from junction-to-ambient. These sources include the junction-to-case resistance, the case-to-heat sink interface resistance, and the heat sink resistance. Thermal resistance specifications have been developed to more accurately reflect device temperature and ensure safe operating temperatures. The electrical characteristics section provides a separate thermal resistance and maximum junction temperature for both the control circuitry and the power transistor. Calculate the maximum junction temperature for both sections to ensure that both thermal limits are met.

For example, look at using an RC1587T to generate 3A @  $1.5V \pm 10\%$  from a 3.3V source (3.2V to 3.6V).

#### **Assumptions:**

- $V_{IN} = 3.6V$  worst case
- Vout = 1.35V worst case
- IQUT = 3A continuous
- $TA = 70^{\circ}C$
- θCase-to-Ambient = 7°C/W (assuming both a heatsink and a thermally conductive material)

The power dissipation in this application is:

$$P_D = (V_{IN} - V_{OUT}) * (I_{OUT}) = (3.6 - 1.35) * (3) = 6.75W$$

The junction will be:

$$TJ = TA + (PD) * (\theta Case-to-Ambient + \theta JC)$$

From the specification table, for the power section:

$$T_J = T_A + (PD) * (\theta_{Case-to-Ambient} + \theta_{JC})$$
  
= 70 + (6.75) \* (7 + 3) = 138°C

For the control section:

$$T_J = T_A + (PD) * (\theta_{Case-to-Ambient} + \theta_{JC})$$
  
= 70 + (6.75) \* (7 + 0.7) = 122°C

In each case, the junction temperature is below the maximum rating for both respective sections.

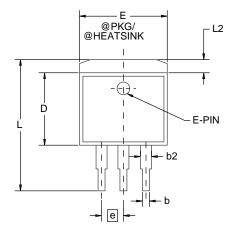
Junction-to-case thermal resistance is specified from the IC junction to the bottom of the case directly below the die. This is the lowest resistance path for heat flow. Proper mounting ensures the best thermal flow from this area of the package to the heat sink. Use of a thermally conductive material at the case-to-heat sink interface is recommended. Use a thermally conductive spacer if the case of the device must be electrically isolated and include its contribution to the total thermal resistance. The case of the RC1587 is directly connected to the output of the device.

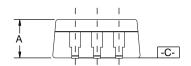
### **Mechanical Dimensions**

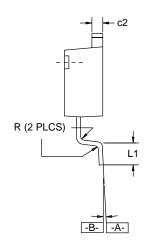
### 3-Lead TO-263 Package

Cumbal	Inches		Millin	Neter	
Symbol	Min.	Max.	Min.	Max.	Notes
Α	.160	.190	4.06	4.83	
b	.020	.039	0.51	0.99	
b2	.049	.051	1.25	1.30	
c2	.045	.055	1.14	1.40	
D	.340	.380	8.64	9.65	
Е	.380	.405	9.65	10.29	
е	.100 BSC		2.54	2.54 BSC	
L	.575	.625	14.61	10.88	
L1	.090	.100	2.29	2.79	
L2	_	.055	_	1.40	
R	.017	.019	0.43	0.48	
α	0°	8°	0°	8°	

- 1. Dimensions are exclusive of mold flash and metal burrs.
- 2. Standooff-height is measured from lead tip with ref. to Datum -B-.
- Foot length ismeasured with ref. to Datum -A- with lead surface (at inner R).
- 4. Dimension exclusive of dambar protrusion or intrusion.
- Formed leads to be planar with respect to one another at seating place -C-.





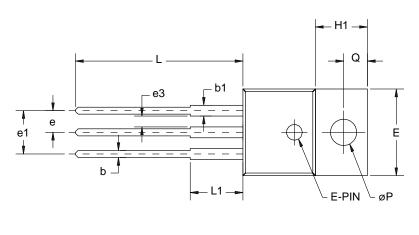


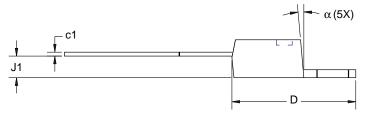
## **Mechanical Dimensions** (continued)

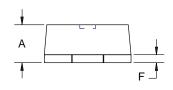
### 3-Lead TO-220 Package

Cumbal	Inc	hes	Millimeters		Notes
Symbol	Min.	Max.	Min.	Max.	Notes
А	.140	.190	3.56	4.83	
b	.015	.040	.38	1.02	
b1	.045	.070	1.14	1.78	
c1	.014	.022	.36	.56	
øΡ	.139	.161	3.53	4.09	
D	.560	.650	14.22	16.51	
E	.380	.420	9.65	10.67	
е	.090	.110	2.29	2.79	
e1	.190	.210	4.83	5.33	
e3	.045		1.14	_	
F	.020	.055	.51	1.40	
H1	.230	.270	5.94	6.87	
J1	.080	.115	2.04	2.92	
L	.500	.580	12.70	14.73	
L1	.250 BSC		6.35 BSC		
Q	.100	.135	2.54	3.43	
α	3°	7°	3°	7°	

Notes:
1. Dimension c1 apply for lead finish.







### **Ordering Information**

Product Number	Package
RC1587M	TO-263
RC1587T	TO-220
RC1587M-1.5	TO-263
RC1587T-1.5	TO-220

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